

# **BJT Modeling With VBIC, Basics and v1.3 Updates**

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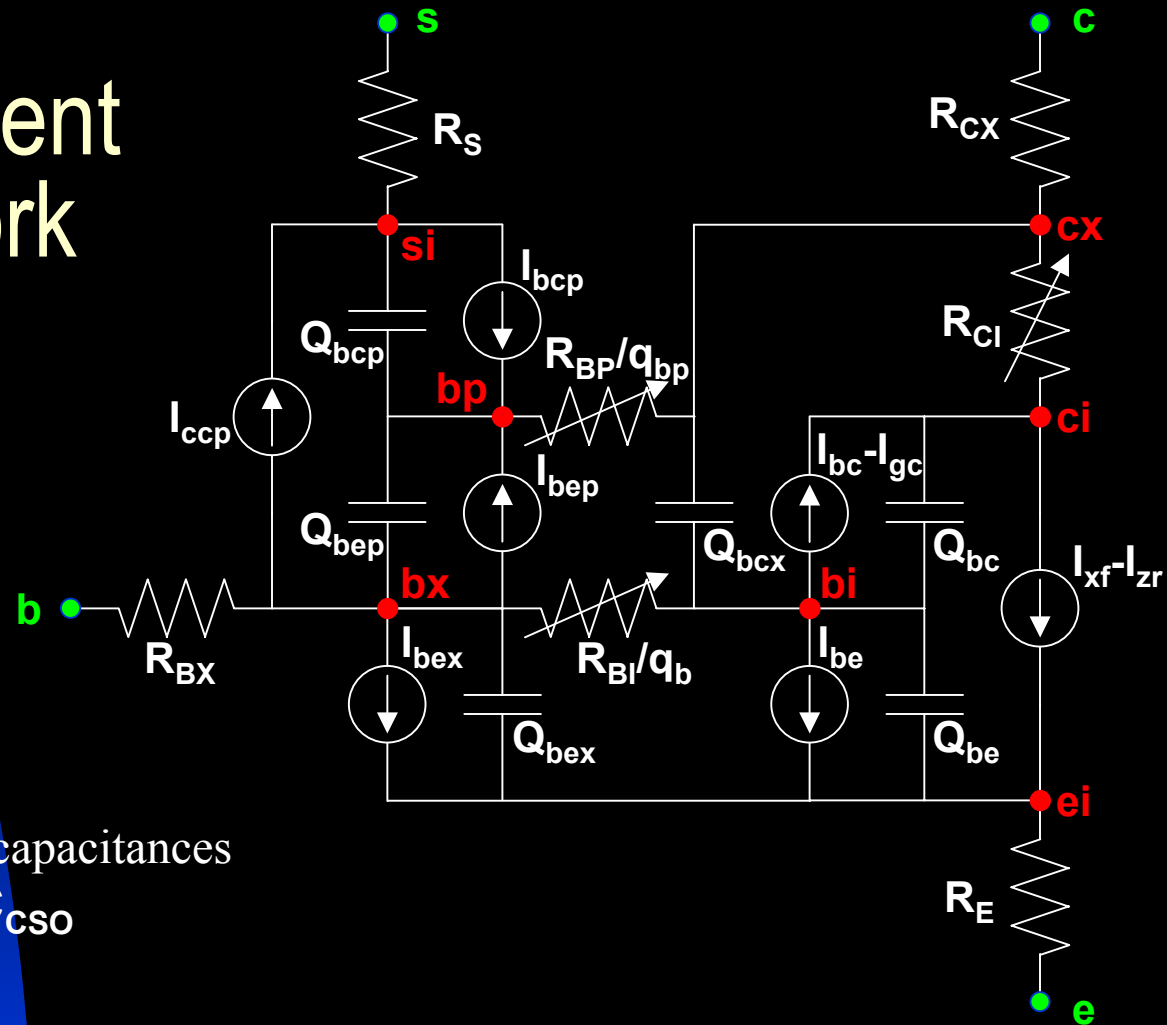


# Outline

- Introduction
- Review of VBIC v1.2
- Updates for VBIC v1.3
  - ◆ General simulator/convergence updates
  - ◆ Other general updates
  - ◆ Depletion  $R_C$  model for vertical PNPs
  - ◆ Excess phase reformulation
  - ◆ Quasi-neutral base recombination
  - ◆ IBM and NSC updates



# Equivalent Network



Not shown:  
Fixed overlap capacitances  
 $C_{BEO}$ ,  $C_{BCO}$ ,  $C_{CSO}$



# VBIC v1.2 Compared to SGP

- Separation of  $I_c$  and  $I_b$ , not linked via  $B_F$
- Early effect model based on depletion charge
- Full parasitic PNP modeling
- Self-heating modeling
- Base-emitter breakdown model
- Weak avalanche model
- Improved depletion, diffusion charge models
- Improved temperature modeling
- Constant overlap capacitances



# VBIC v1.3

- Under development now
- Some updates finished, others planned
- General areas of enhancement
  - ◆ Robustness
  - ◆ Incremental additions (e.g. extra TCs)
  - ◆ Improvements for higher voltage device modeling
  - ◆ Other additions (IBM, NSC)



# Robustness

- `gmin` and `pnjmaxi` explicitly added to model formulation
  - ◆ Before expected to be added, simulator specific
- Explicit linearization added, above `pnjmaxi`
- Improved numerical protection for normalized base charge and weak avalanche models
- Max/Min implemented for local temperature
- Shrink factor support added



# PNPV $R_C$ Modulation

- No P<sup>+</sup> buried layer
- Depletion pinching of collector, by substrate/isolation, modulates  $R_C$
- Can be modeled, approximately, in a subckt by including a JFET for  $R_C$
- Included as alternative  $R_C$  modulation model, based on diffused resistor model

$$I_{rci} = \frac{V_{rci}}{R_{CI}} \left( 1 - D_{FC} \sqrt{P_{FC} + 2V_{cx-si} + V_{rci}} \right)$$



# Base-Collector “Pinning”

- For lightly doped collectors with no buried layer, the base-collector depletion region can push through to the substrate/isolation
- $C_{bc}$  drops sharply at that point
- Early voltage increases sharply at that point
  - ◆ Still some Early effect from sidewall
- Implemented via two components of  $Q_{BC}$ , with a “pinning” voltage at which the charge from one component (smoothly) limits



# Self-Heating Modeling

- VBIC was the first new generation BJT model to include self-heating
  - ◆ Primarily because of extremely easy implementation, automated code generation from a high-level model description
- Simple single  $R_{TH}$  and  $C_{TH}$  model, also adopted by other models, has limitations
- VBIC v1.3 has  $R_{TH}(T)$  model
- Considering adding multi-pole  $Y_{TH}$  model



VBIC MODEL APPLICABILITY AND EXTRACTION PROCEDURE FOR  
InGaP/GaAs HBT

S.V. CHEREPKO and J.C.M HWANG  
Lehigh University

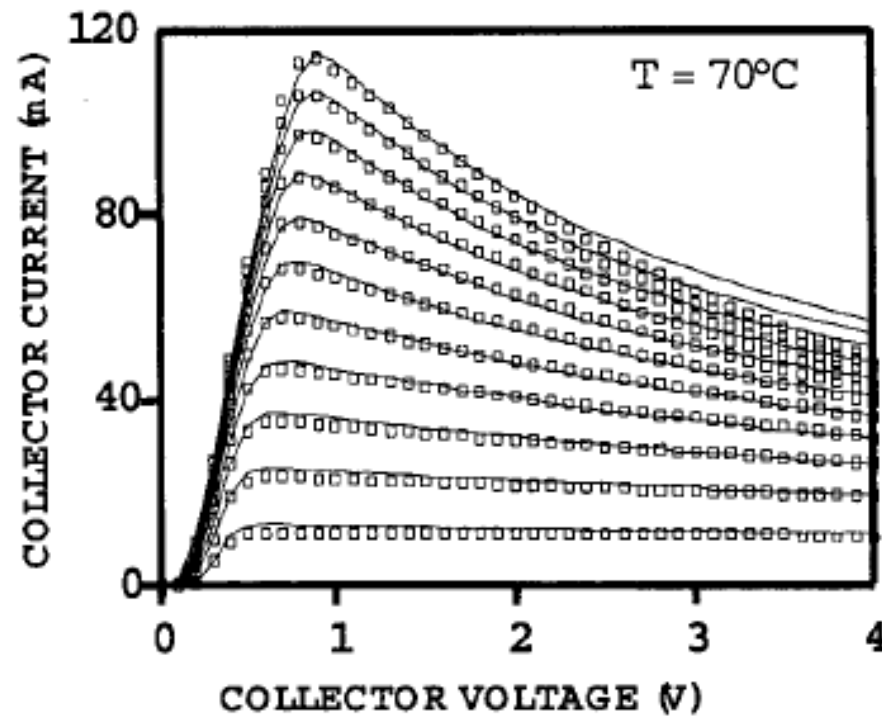


Fig. 4 (symbols) Measured vs. (lines) simulated  $I$ - $V$  characteristics.  
 $I_B = 0.075, 0.15, \dots, 0.875$  mA bottom top.



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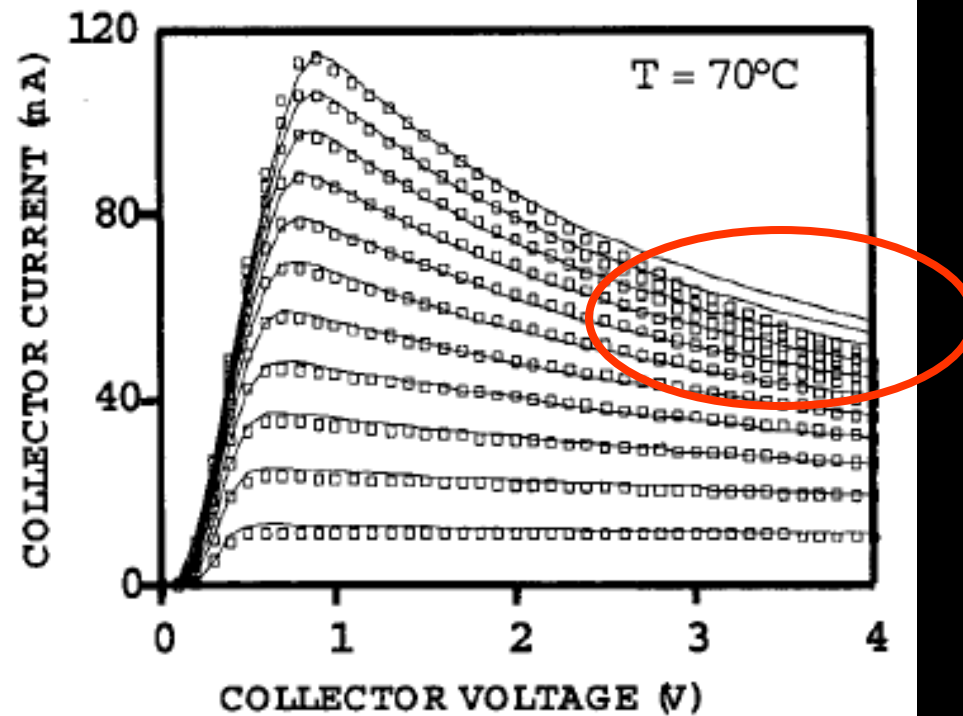
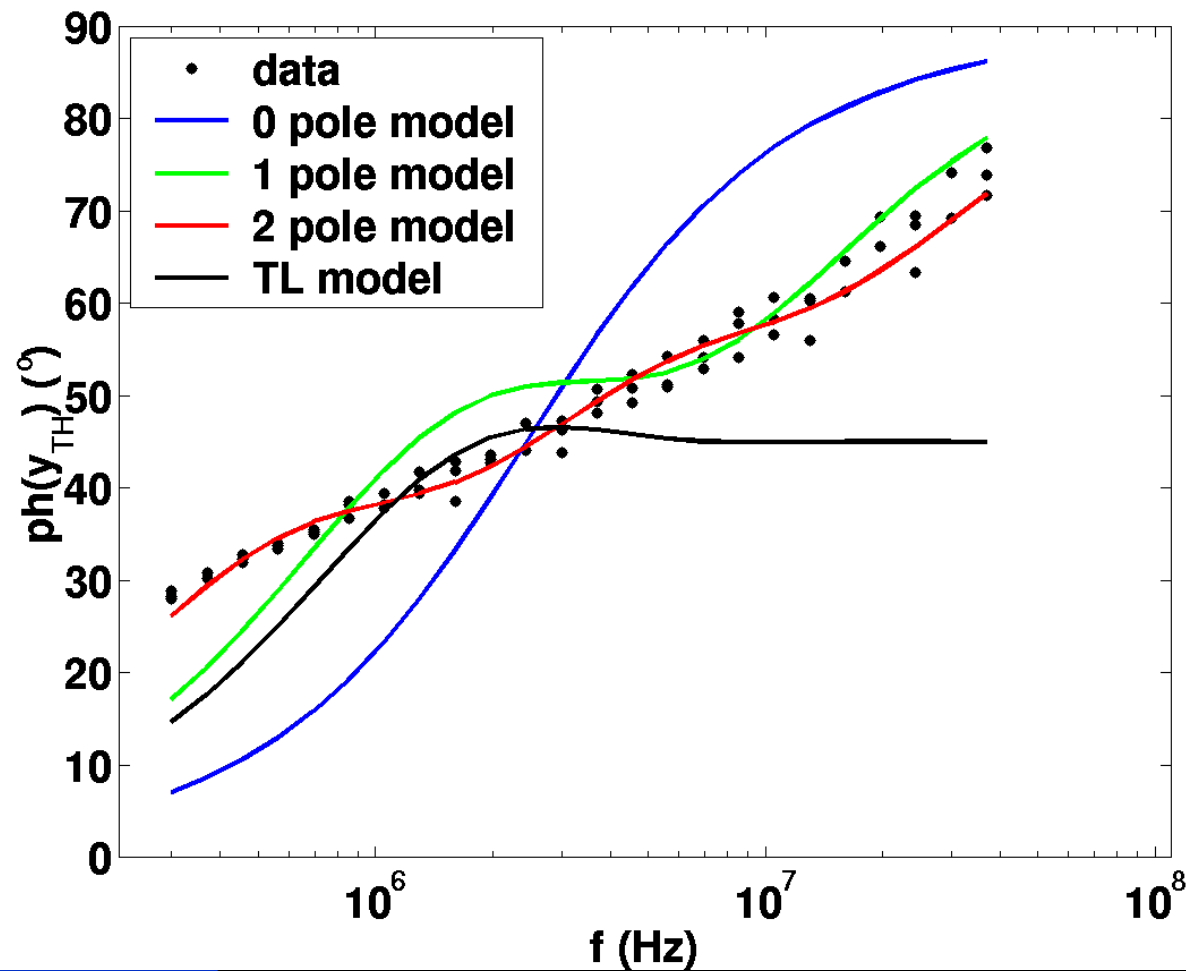


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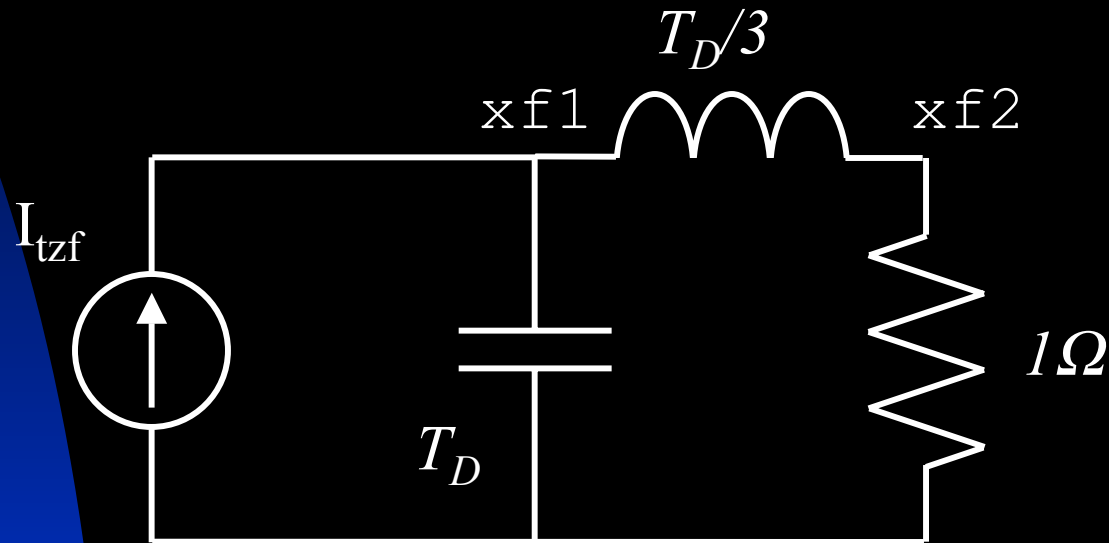


# Extracted $\text{Ph}(y_{\text{TH}})$



# Excess-Phase Remapping

- Original Weil-McNamee formulation implemented as RLC circuit



# Excess-Phase Remapping

- Consistent for all analyses types, not just for AC or transient
- Inductor is messy to handle in nodal formulation
- Only really 2 unknowns as inductor current equals  $V(xf2)$ , do not have to carry extra current in MNA formulation
- Non-standard for implementation



# Excess-Phase Remapping

$$I_{tzf} - ddt(T_D V(xf1)) - I_L = 0$$

$$I_L - V(xf2) = 0$$

$$V(xf1) - V(xf2) = ddt(I_L T_D / 3)$$



# Excess-Phase Remapping

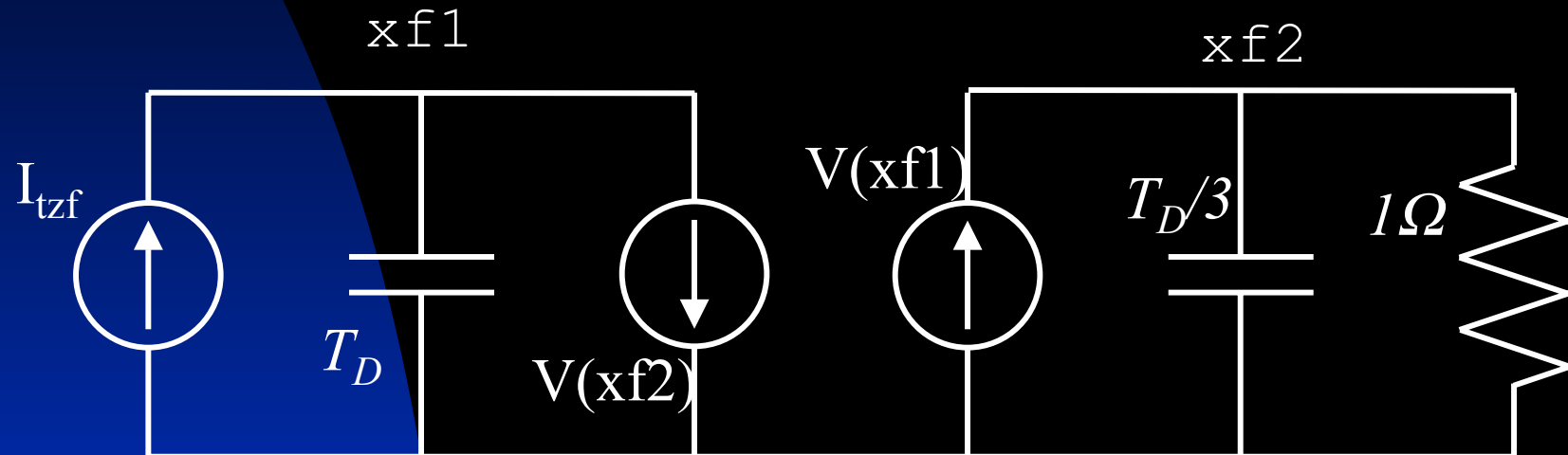
$$I_{tzf} - ddt(T_D V(xf1)) - V(xf2) = 0$$

$$V(xf2) + ddt(V(xf2)T_D/3) - V(xf1) = 0$$



# Excess-Phase Remapping

- Equivalent in VCCS type elements



# Early Voltages

- Temperature dependence of the Early effect determined by the temperature dependence of depletion charge model
- Not accurate enough for some devices, especially those with low built-in potentials at high temperature
- Linear TCs added for Early voltages

$$V_{EF}(T) = V_{EF}(T_{NOM})(1 + TC_{VEF}(T - T_{NOM}))$$



# Reverse Operation

- First order theory ties, through reciprocity, reverse  $I_S$  to forward  $I_S$
- Data shows this is not always a good approximation
  - ◆ Especially for III-V HBTs
- A separate  $I_S$  is provided for reverse operation, with a separate TC
- Implemented as ratio w.r.t. forward  $I_S$ , for simple extraction and compatibility



# Transit Time

- VBIC v1.2 transit time model from SGP, with minor addition of depletion modulation
- Recognized weak part of the model
- Is being updated for v1.3, based on both SiGe HBTs and high voltage devices (NPN and PNP) in SmartPower technologies

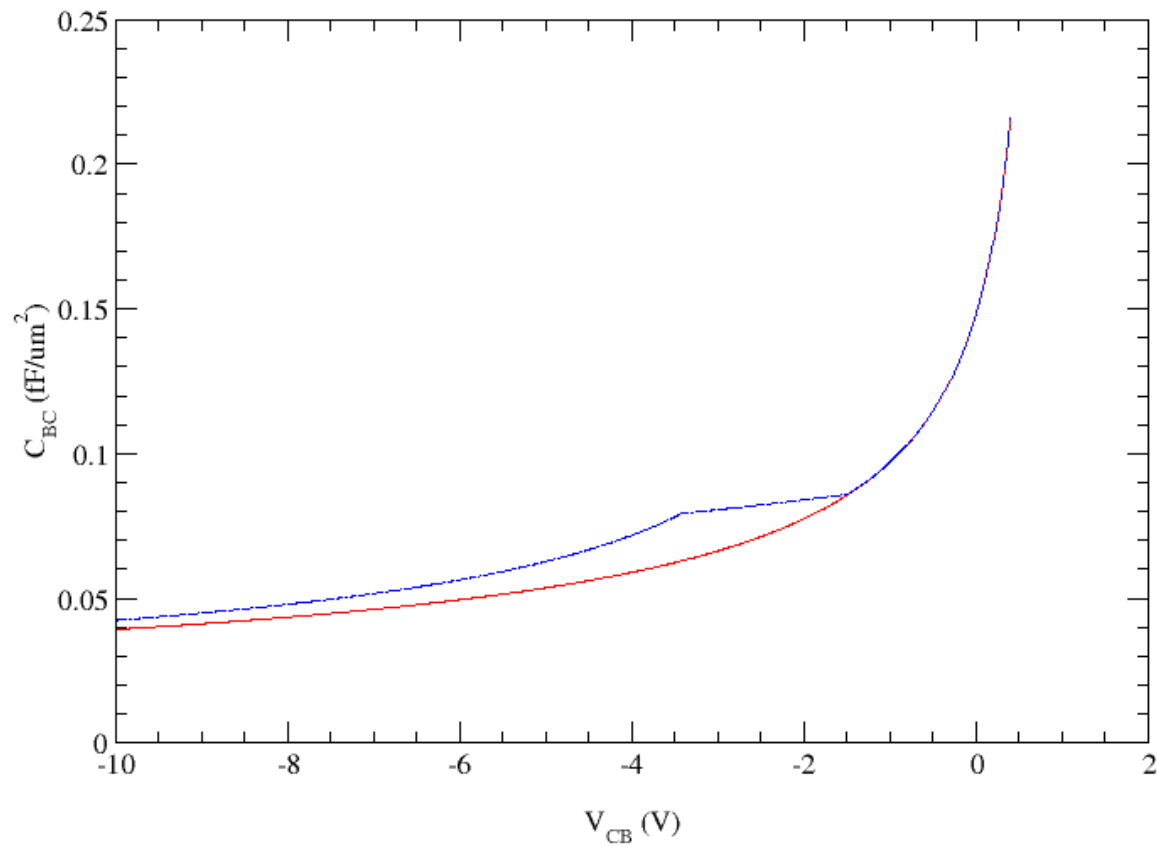


# Variable Collector Doping Model

- Has been shown that adjusting the collector doping profile with different levels and spikes can improve large signal linearity
- Some devices also have variable dopings in the collector
- Empirical collector doping model implemented
  - ◆ Parameterized in terms of voltage rather than depth, for simple parameter extraction



# Collector-Base Capacitance



# Quasi-Neutral Base Recombination

- VBIC  $I_b$  model is based on recombination and generation in the quasi-neutral emitter, and the emitter contact, and in the base-emitter space-charge region
- In wide base devices neutral base recombination also occurs
  - ◆ Visible in output characteristics, and in  $\beta$  variation with collector bias



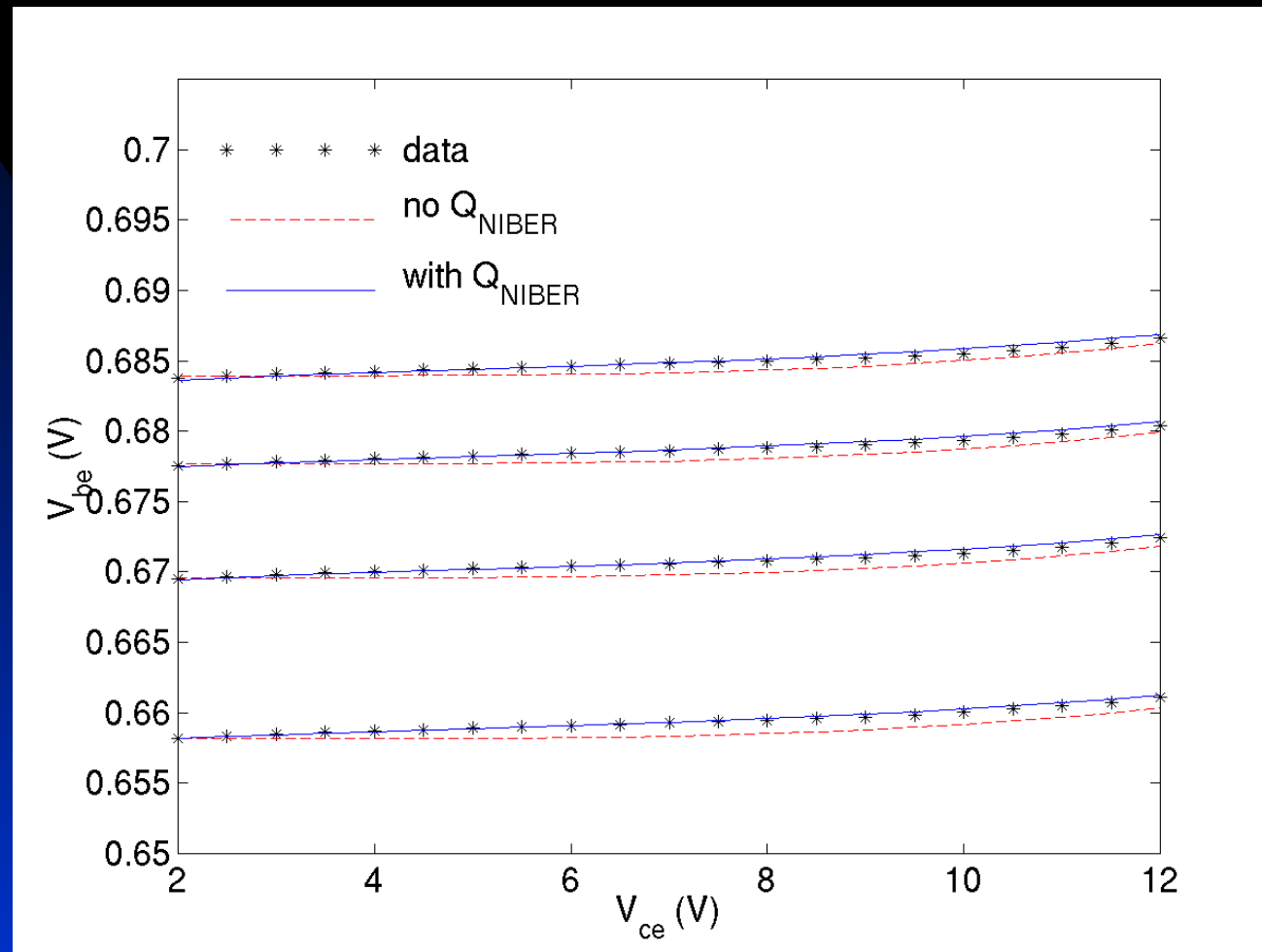
# Quasi-Neutral Base Recombination

- The base volume is linked to the  $q_1$  component of normalized base charge
  - ◆ Although details of  $N(x)$ , which affects R/G, are not included
- Separate  $I_{BEI}$  into emitter and base components

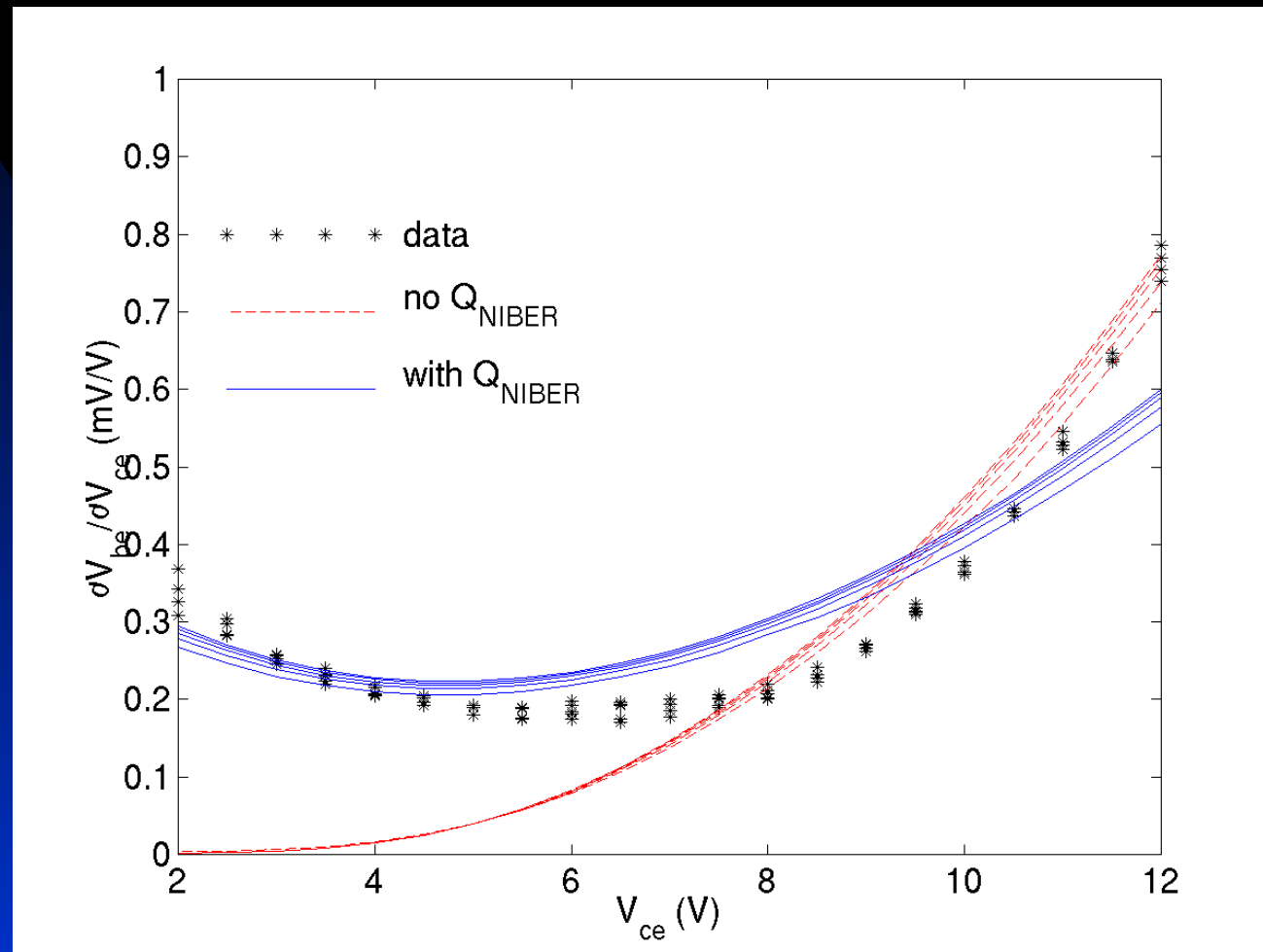
$$I_{BEI} = I_{BEI}^e + I_{BEI}^b q_1 = I_{BEI}^0 (1 + Q_{NIBEIR} (q_1 - 1))$$



# Quasi-Neutral Base Recombination



# Quasi-Neutral Base Recombination



# IBM Updates

- Alternative placement of control voltages for weak avalanche for high breakdown devices
- Slightly modified avalanche expression for this also
- Separate depletion capacitance parameters for  $C_{JC}$  and  $C_{JEP}$
- Temperature dependent transit times
  - ◆ Linear temperature dependence
- Acknowledge D. Sheridan for provision of details



# NSC Updates

- Model for  $I_B$  bias and geometry dependence for special effect seen in SiGe devices provided by A. Sadovnikov of NSC
- Details to be published elsewhere



# Summary

- Updates for VBIC v1.3 in progress at present, target release date April 2003
- Details of some updates provided
- Others still in implementation
- Model is defined in Verilog-A, updates are implemented there, C code is generated from this

