

Theory, development, and applications of the Advanced Compact MOSFET (ACM) model

**Carlos Galup-Montoro, Márcio C. Schneider,
Ana I. A. Cunha*, and Oscar C. Gouveia Filho****

Federal University of Santa Catarina

****Federal University of Bahia***

*****Federal University of Paraná***

Brazil

OUTLINE

- Introduction
- Fundamentals
- The Unified Charge Control Model
- Drain Current
- Equations for Circuit Design
- Parameter Extraction
- Short-Channel Effects
- ACM Model for Simulators
- Summary

Introduction

Genealogy of the ACM Model

Charge-sheet model (Brews – 78)

Charge-based models

Maher & Mead – 87

UCCM – 90

Iñiguez & Moreno 93

Symmetric models

Bentchkowsky & Vadasz - 69

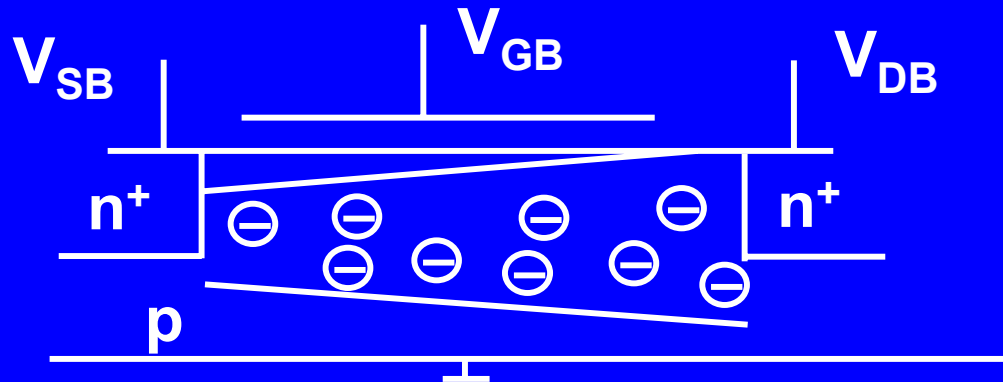
Lo & Gibson 73

EKV - 89

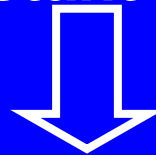
ACM

Fundamentals

Depletion-Capacitance Approximation



C'_b , calculated assuming the inversion charge density to be negligible, is constant along the channel



$$Q_I' = -C_{ox}' \left(V_{GB} - V_{FB} - \phi_s - \gamma \sqrt{\phi_s} \right) \cong \left(1 + \frac{\gamma}{2\sqrt{\phi_{sa}}} \right) C_{ox}' (\phi_s - \phi_{sa}) = (C_{ox}' + C_b') (\phi_s - \phi_{sa})$$

Basic relationships:

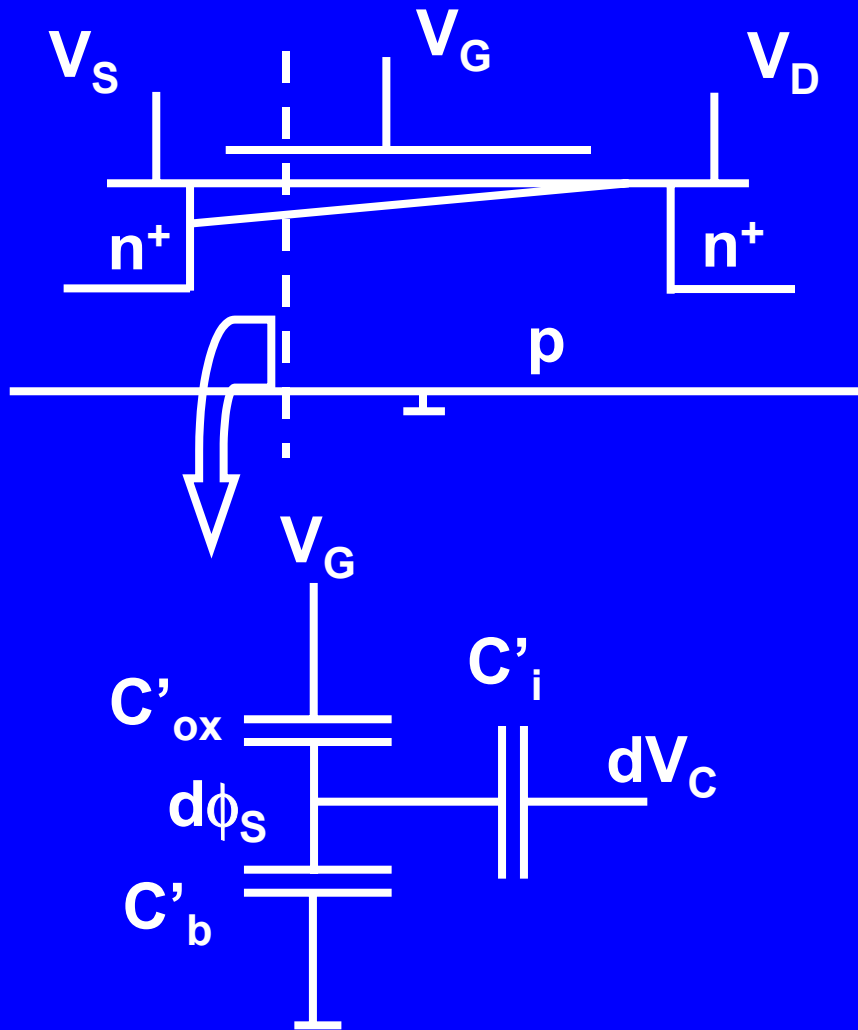
The inversion and depletion charge densities are incrementally linear functions of the surface potential

$$dQ'_I = (C'_{ox} + C'_b) d\phi_s = nC'_{ox} d\phi_s$$

$$dQ'_B = -C'_b d\phi_s = -(n-1)C'_{ox} d\phi_s$$

$$C'_b = C'_{ox} \frac{\gamma}{2\sqrt{\phi_{sa}}}$$

The Unified Charge Control Model



$$\left. \frac{dQ'_I}{dV_C} \right|_{V_G} = \frac{(C'_{ox} + C'_b)C'_i}{C'_{ox} + C'_b + C'_i}$$

Basic approximations:

$$C'_{ox} + C'_b = nC'_{ox}$$

$$C'_i = -Q'_I / \phi_t$$

$$dQ'_I \left(\frac{1}{nC'_{ox}} - \frac{\phi_t}{Q'_I} \right) = dV_C$$

The Unified Charge Control Model

Integrating $dQ'_i \left(\frac{1}{nC'_{ox}} - \frac{\phi_t}{Q'_i} \right) = dV_C$ between V_C and V_P :

$$\boxed{\frac{Q'_{IP} - Q'_i}{nC'_{ox}} + \phi_t \ln \left(\frac{Q'_i}{Q'_{IP}} \right) = V_P - V_C} \quad \text{UCCM}$$

$Q'_{IP} = -nC'_{ox}\phi_t$: inversion charge at “pinch-off”

In strong inversion: $-Q'_i \cong nC'_{ox}(V_P - V_C)$

In weak inversion: $\phi_t \ln \left(\frac{Q'_i}{Q'_{IP}} \right) \cong V_P - V_C$

Drain Current

$$I_D = \mu W \left(-Q'_I \frac{d\phi_s}{dx} + \phi_t \frac{dQ'_I}{dx} \right)$$

↓ drift ↓ diffusion

Using the approximation

$$dQ'_I = n C'_{ox} d\phi_s$$

and integrating from source to drain:

$$I_D = -\frac{\mu}{n C'_{ox}} \frac{W}{L} \int_{Q'_{IS}}^{Q'_{ID}} (Q'_I - n C'_{ox} \phi_t) dQ'_I$$

$$I_D = I_F - I_R = I(V_G, V_S) - I(V_G, V_D)$$

$$I_{F(R)} = \mu n C'_{ox} \frac{W}{L} \frac{\phi_t^2}{2} \left[\left(\frac{Q'_{IS(D)}}{n C'_{ox} \phi_t} \right)^2 - 2 \frac{Q'_{IS(D)}}{n C'_{ox} \phi_t} \right]$$

Drain Current

$$\frac{Q'_{IS(D)}}{-nC'_{ox}\phi_t} = \sqrt{1 + i_{f(r)}} - 1$$

$$i_{f(r)} = \frac{I_{F(R)}}{I_S}$$

Forward (reverse) normalized current

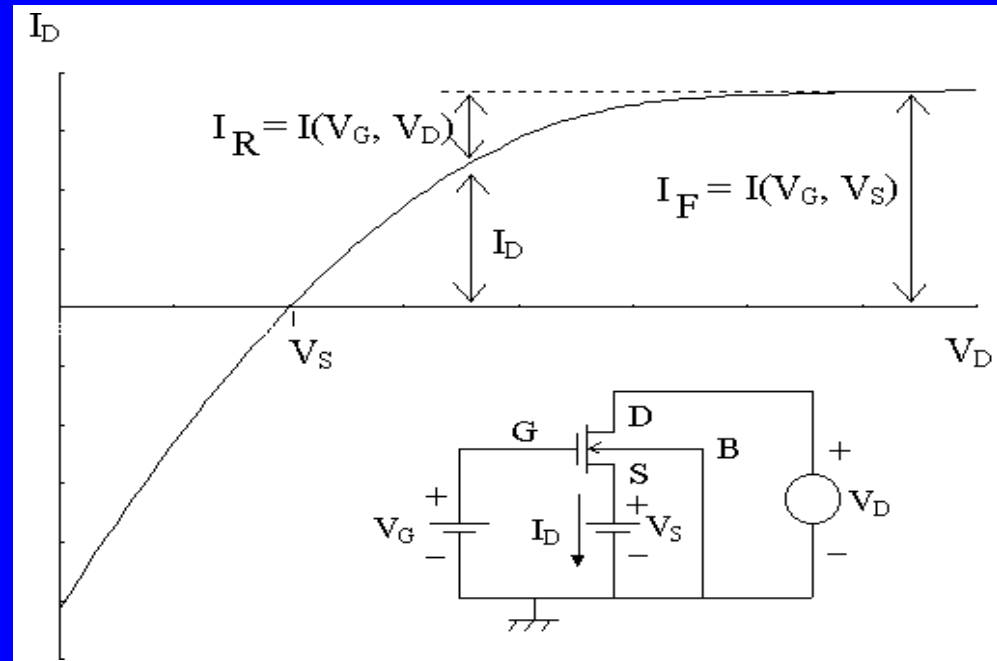
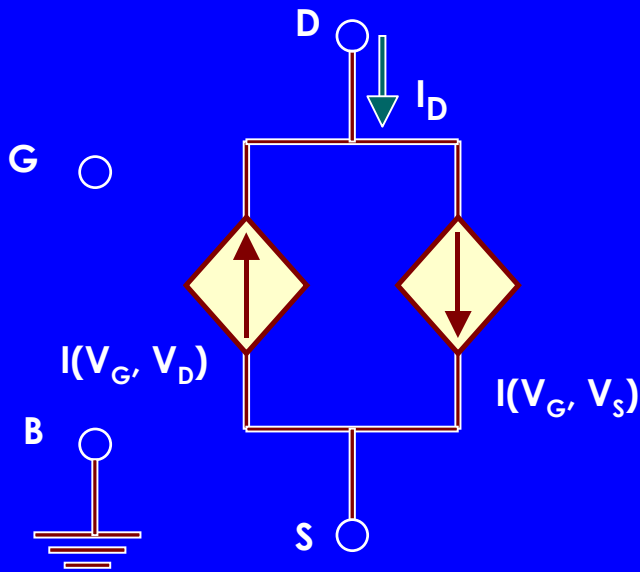
$$I_S = \frac{1}{2} n \mu C'_{ox} \phi_t^2 \frac{W}{L}$$

Normalization (specific) current

UICM \longrightarrow
$$\sqrt{1 + i_{f(r)}} - 2 + \ln\left(\sqrt{1 + i_{f(r)}} - 1\right) = \frac{V_P - V_{S(D)}}{\phi_t}$$

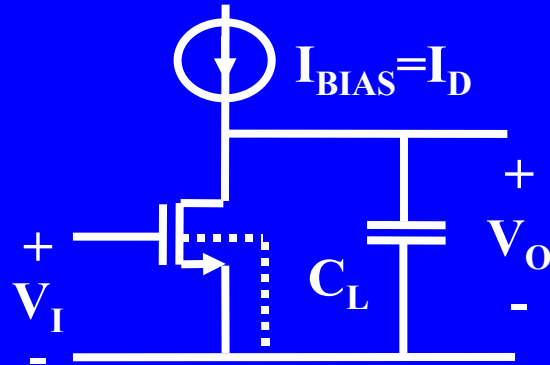
Drain Current: Forward and Reverse Currents

$$I_D = I_F - I_R = I(V_G, V_S) - I(V_G, V_D)$$



For large V_D , $I_R \ll I_F$; therefore, $I_D \cong I_F$

Equations for Circuit Design

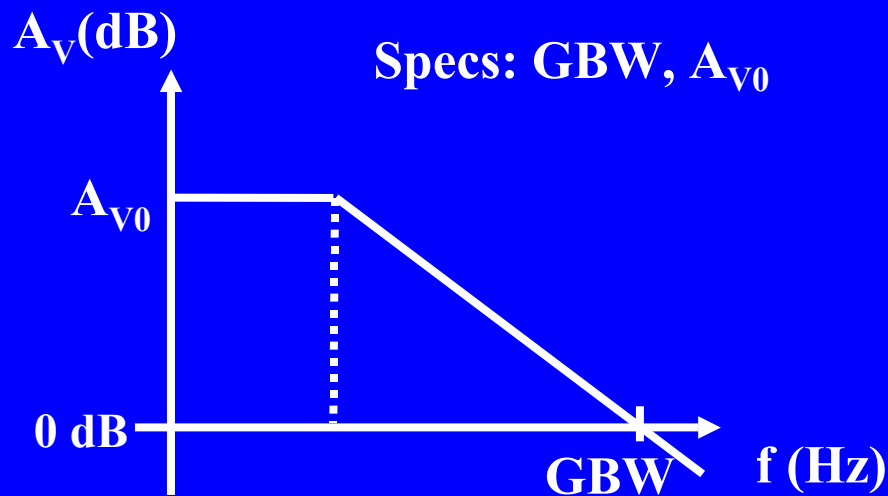


Design parameters $\rightarrow I_D, W, L$

$$I_D = n\phi_t g_m \frac{1 + \sqrt{1 + i_f}}{2}$$

$$L = A_{V0} \frac{n\phi_t}{V_E} \frac{1 + \sqrt{1 + i_f}}{2}$$

$$\frac{W}{L} = \frac{g_m}{\mu C'_{ox} \phi_t} \frac{1}{\sqrt{1 + i_f} - 1}$$

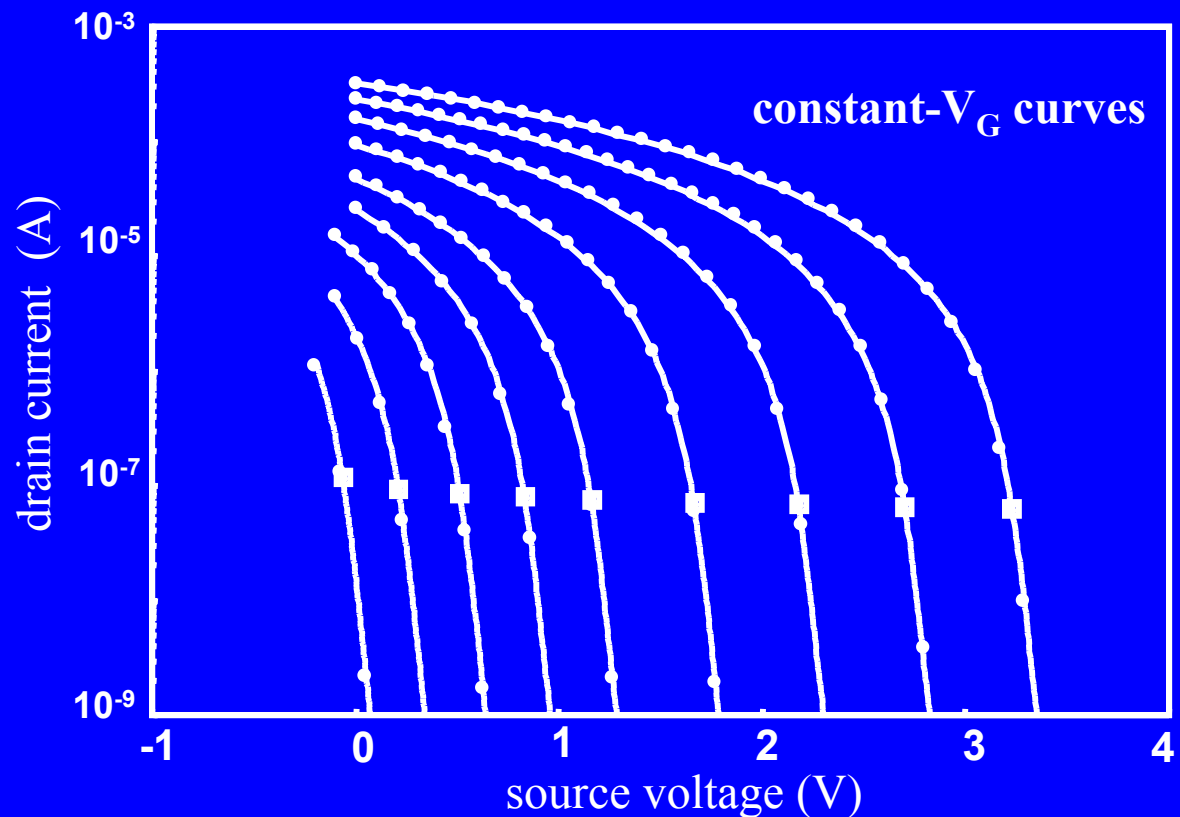
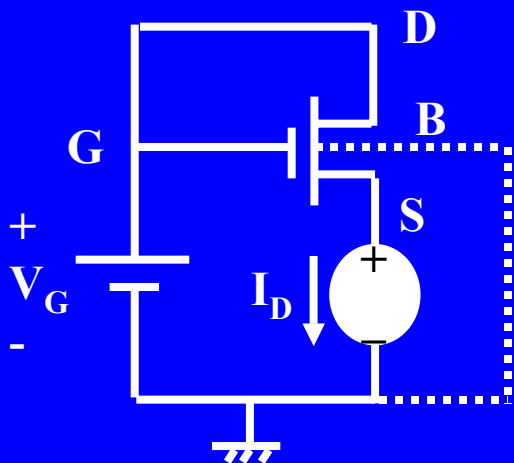


$$GBW = g_m / 2\pi C_L$$

$$A_{V0} = g_m / g_o$$

Parameter Extraction

Principle for the extraction of I_S , V_T , and $n \rightarrow$ The percent variation of the drain current wrt the source voltage depends on $i_f = I_D/I_S$



Short-Channel Effects

- **DIBL : drain-induced barrier lowering**

$$I_D = f(V_P, V_S) - f(V_P, V_D)$$

$$V_P(V_G, V_S, V_D) = V_{P0}(V_G) + \frac{\sigma}{n}(V_D + V_S)$$

$V_{P0} \Rightarrow$ **long-channel pinch-off voltage**

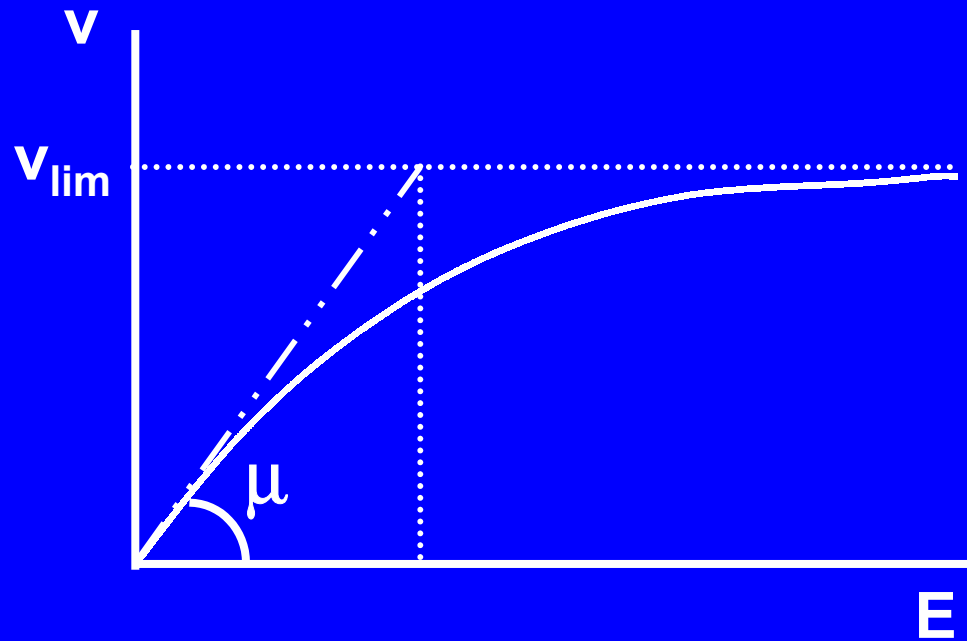
$\sigma \Rightarrow$ **fitting parameter (depends on technology and L)**

• Velocity saturation

$$\mu_s = \frac{\mu}{1 + \frac{\mu}{v_{lim}} \frac{d\phi_s}{dx}}$$

$$\frac{d\phi_s}{dx} = -E \quad (\text{longitudinal field})$$

$$v_{lim} = 5 \times 10^6 \text{ to } 2 \times 10^7 \text{ cm/s}$$

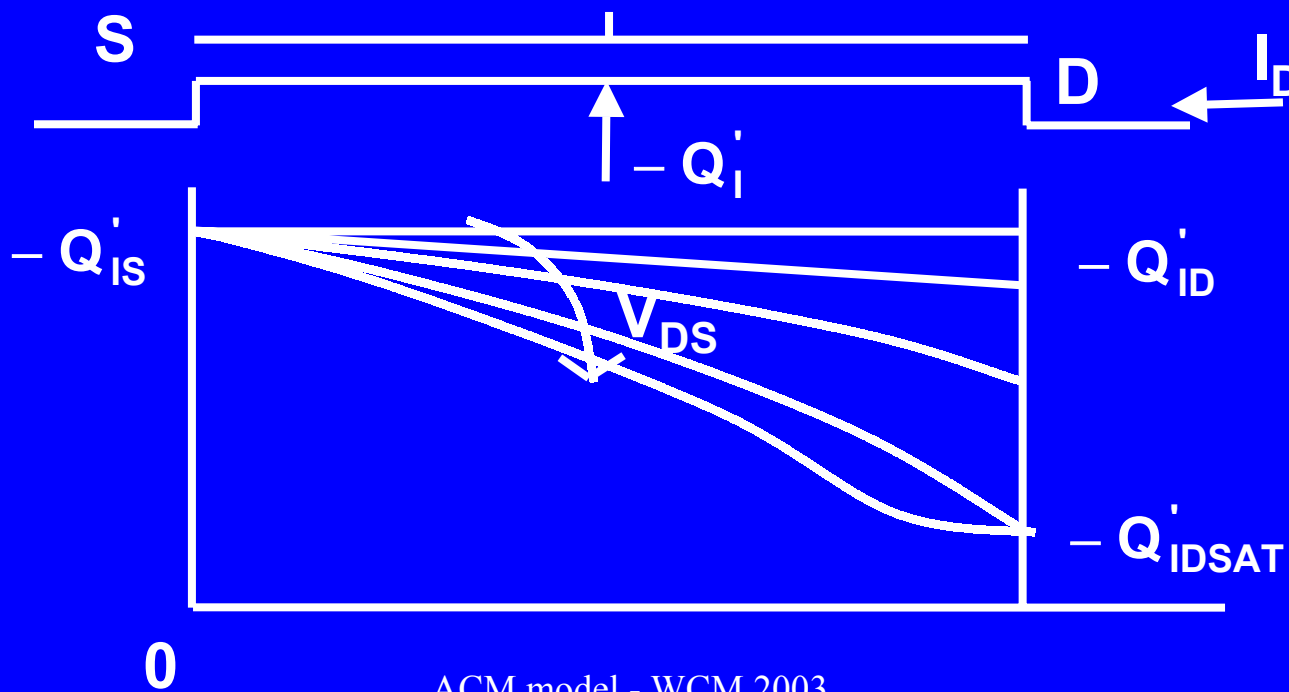


•Velocity saturation (continued)

Saturation:

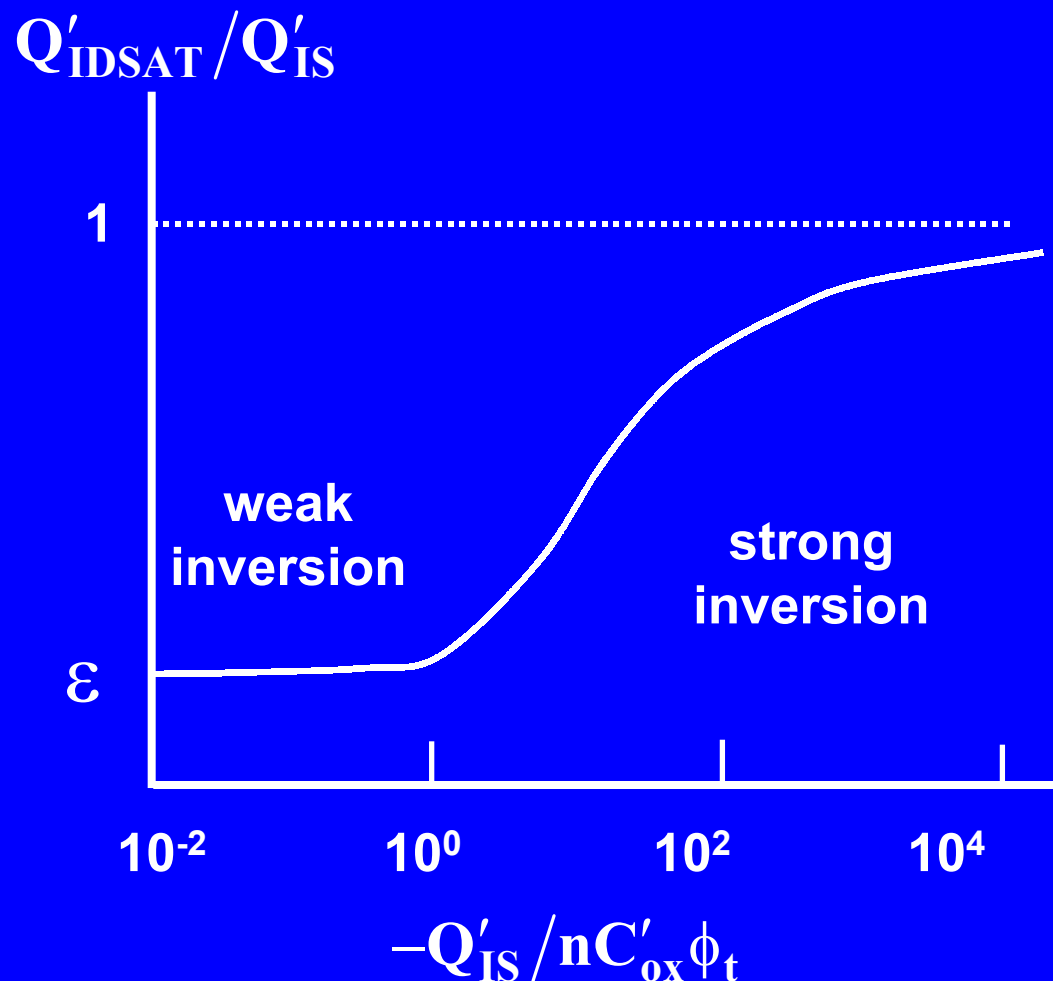
The minimum amount of electron charge flowing at the saturation velocity required to sustain the current is:

$$Q'_{IDSAT} = -I_D / Wv_{lim}$$



•Velocity saturation (continued)

$$\varepsilon = \frac{\phi_t / L}{v_{lim} / \mu}$$



•Effects of DIBL and velocity saturation on intrinsic charges

V_p includes DIBL $\rightarrow Q'_{IS(D)}$ includes DIBL

$$\text{charge} = f(Q'_{IS}, Q'_{ID})$$

Drift

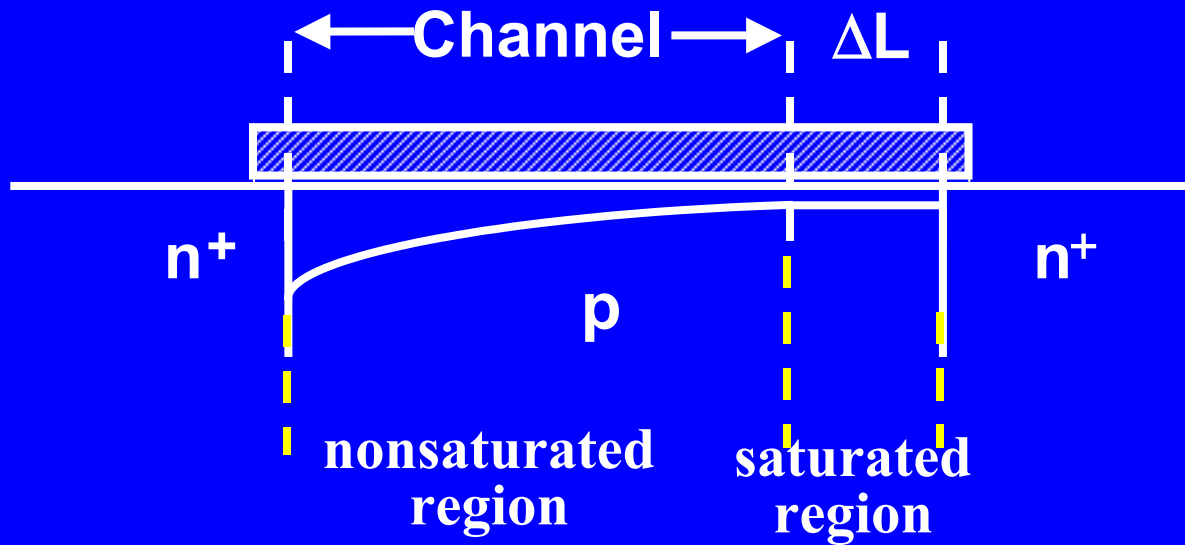
$$\text{charge} = f(Q'_{IS} - nC'_{ox}\phi_t, \dots)$$

Drift and diffusion

$$\text{charge} = f\left(Q'_{IS} - nC'_{ox}\phi_t + \frac{I_D}{W_{\text{eff}} v_{\text{lim}}}, \dots\right)$$

Drift, diffusion and velocity saturation

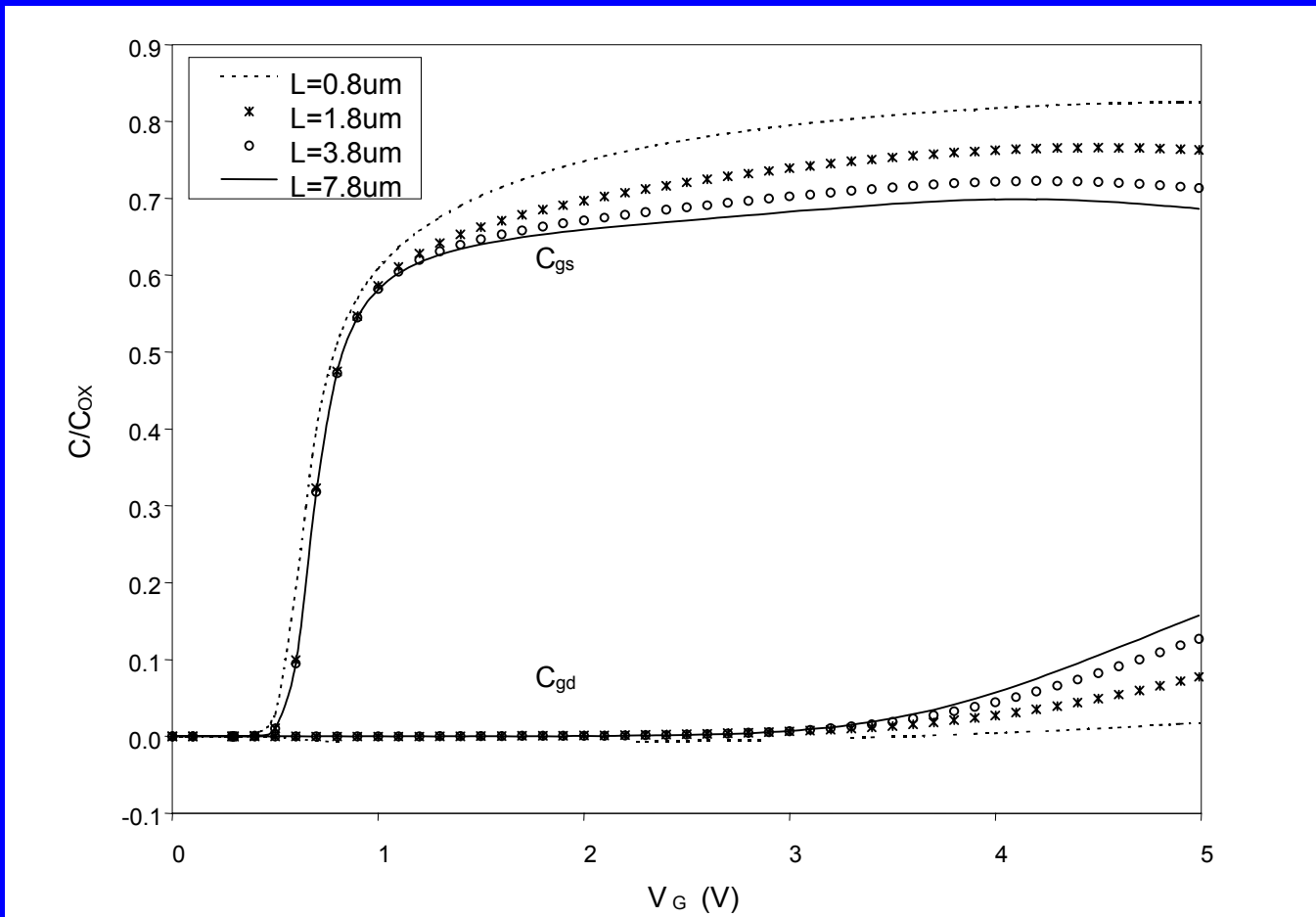
•Short-channel effects on intrinsic charges



$$Q_I = W_{\text{eff}} \int_0^{L_{\text{eq}}} Q'_I dx + W_{\text{eff}} \Delta L Q'_{\text{IDSAT}}$$

$$Q_I = W_{\text{eff}} L_{\text{eff}} \left(\frac{2 Q'_F{}^2 + Q'_F Q'_R + Q'_R{}^2}{3 (Q'_F + Q'_R)} - Q'_{\text{IP}} + Q'_{\text{IDSAT}} \right) + W_{\text{eff}} \Delta L Q'_{\text{IDSAT}}$$

$$Q'_{\text{F(R)}} = Q'_{\text{IS(D)}} + Q'_{\text{IP}} - Q'_{\text{IDSAT}}$$



C_{gs} and C_{gd} versus V_G for channel-lengths varying from $0.8\mu m$ to $7.8\mu m$. $V_D=5V$ and $V_S=V_B=0V$.

ACM Model for Simulators

Parameter	Description	Unit
VTO	Zero-bias threshold voltage	V
GAMMA	Body-effect parameter	$V^{1/2}$
PHI	Surface potential	V
TOX	Gate oxide thickness	m
LD	Lateral diffusion	m
XJ	Junction depth	m
UO	Low-field mobility	cm^2/Vs
VMAX	Saturation velocity	m/s
THETA	Mobility reduction parameter	V^{-1}
SIGMA	DIBL parameter	m^2
PCLM	CLM parameter	-

The ACM model is available in SMASH, a circuit simulator of Dolphin Integration

Summary

Properties of the ACM model:

- MOSFET model based on physics
- Appropriate normalization variables that represent the physics behind the model

Normalization variable	Symbol	Definition
Thermal voltage	ϕ_t	$k_B T/q$
Pinch-off charge density	Q'_{IP}	$-nC'_{ox}\phi_t$
Specific current	I_S	$n\mu C'_{ox}(\phi_t^2/2)(W/L)$
Channel length	L	L
Transit time	τ	$L^2/(2\mu\phi_t)$

Summary (continued)

- Single-piece expressions with infinite order of continuity
- Preserves the MOSFET symmetry
- Conserves charge
- Reduced number of parameters
- Implemented in SMASH, a mixed-mode simulator